

Low-noise Receiver / Scope Combinations For PDV

Kirk Miller, Kevin Lee, Ed Daykin, Jared Gordon,
Tom Waltman, Mike Grover, Jerry Stevens, Brandon LaLone,
Dale Turley, Ben Valencia, Carlos Perez, Abel Diaz, Ted Strand,
Adam Iverson, Sonny Gallegos

National Security Technologies, LLC

Mike Furlanetto, David Holtkamp,
Mitzi Boswell, Benjie Stone, Lenny Tabaka

Los Alamos National Laboratory

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Background

- Currently using high-gain receivers with fast digitizers
- With new generation of digitizers with useful sensitivity below 20 mV/div, there is an opportunity to use lower-gain receivers
- What PDV requires is measurement of multiple frequencies across a very wide variety of signal levels: High dynamic range
- What PDV does not require:
 - accurate phase across long records
 - exquisitely low harmonic distortion
- We will use SNR as a means to estimate the frequency-domain effective number of bits (“FNOB”) across a wide array of input signal levels

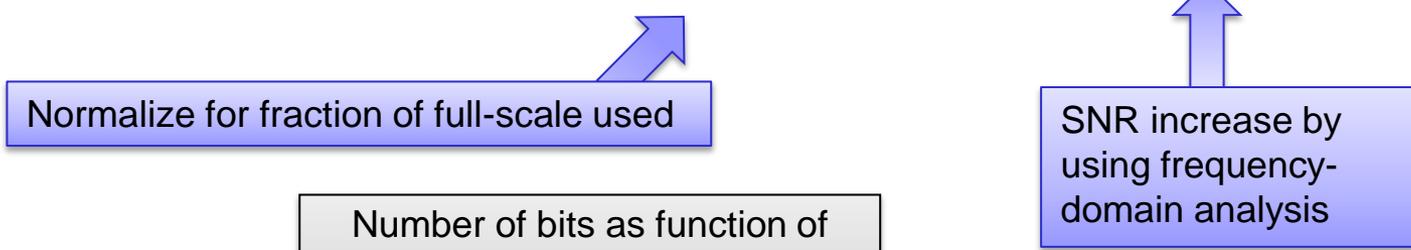


SNR relationship to ENOB

$$SNR_{\text{TIME-DOMAIN,dB}} = 6.02 * ENOB + 1.76 + 20 \log (2 * RMS / V_{\text{full-scale}})$$

ENOB = effective bits for digitizer, V = full scale range, A = RMS amplitude of applied signal
 See Wiley Encyclopedia of Electrical and Electronics Engineering, Vol. 18, J. Blair

$$SNR_{f,\text{dB}} = 6.02 * ENOB + 1.76 + 20 * \log(2 * RMS / V_{FS}) + 10 * \log(N_{\text{FFT}} / 2)$$



“Frequency-domain Number of Bits”

Number of bits as function of SNR, fraction of full-scale, FFT RBW

$$“FNOB” = (1/6.02) * (SNR_{f,\text{dB}} - 1.76 - 20 * \log(2 * RMS / V_{FS}) - 10 * \log(N_{\text{FFT}} / 2))$$

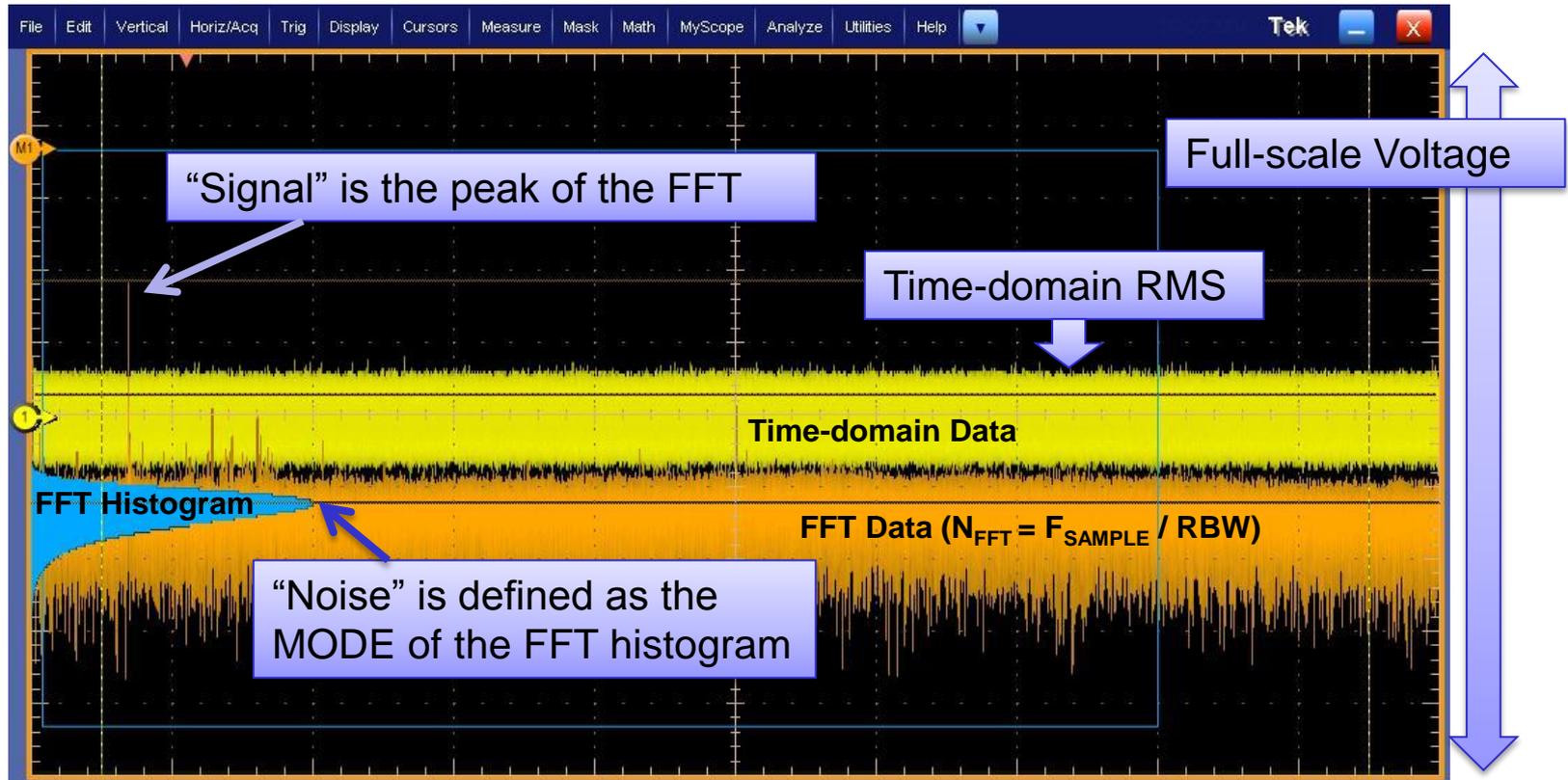
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Measure this

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Measure this

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Fix this using Resolution Bandwidth (RBW) for FFT



On-board analysis for Lab data



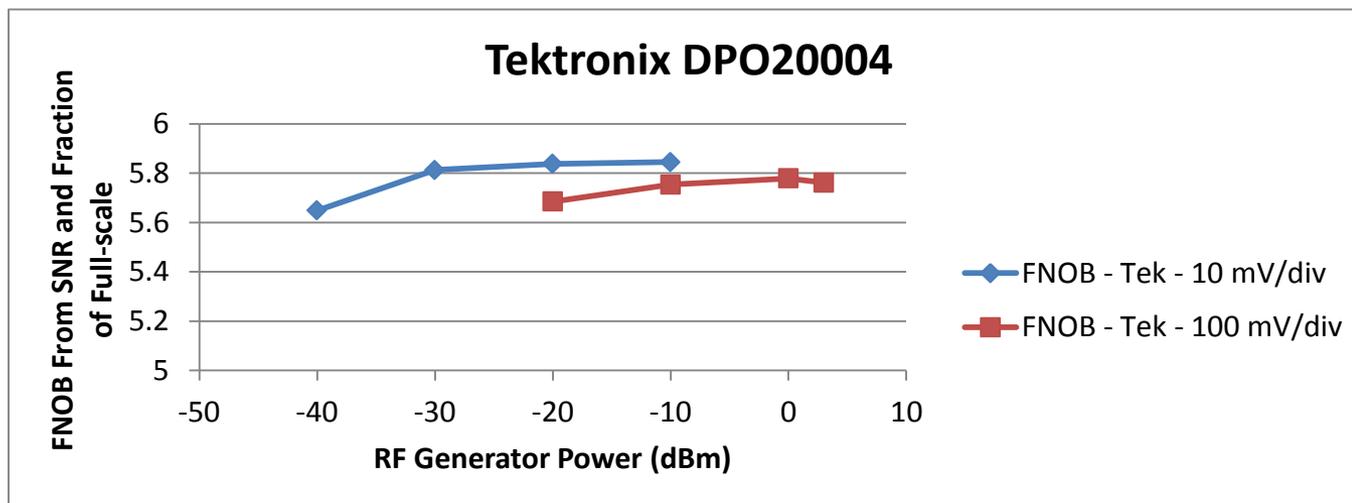
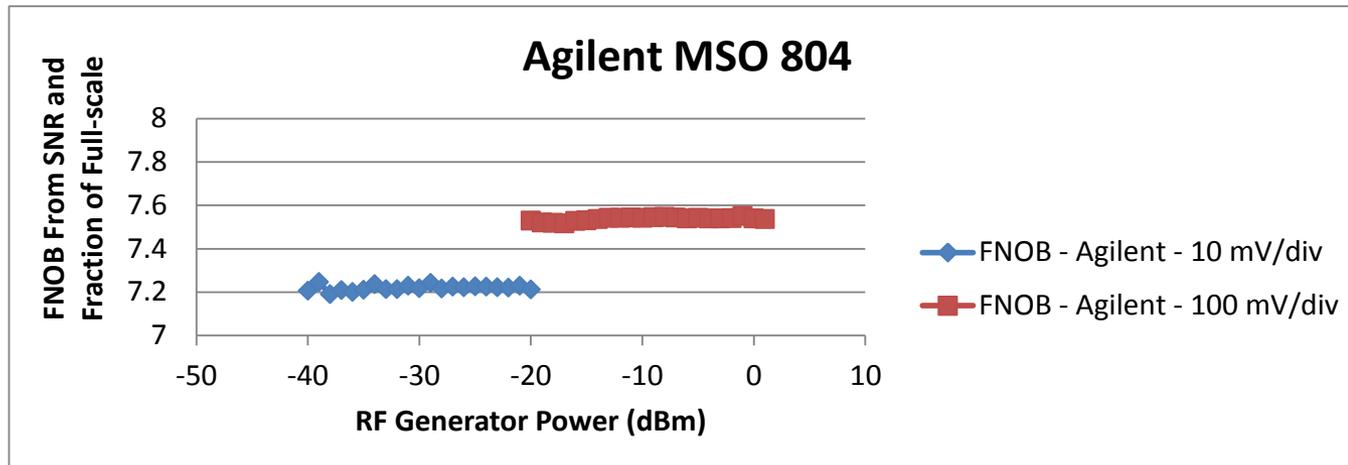
$$\text{“FNOB”} = (1/6.02) * (\text{SNR}_{f,\text{dB}} - 1.76 - 20 * \log(2 * \text{RMS} / V_{FS}) - 10 * \log(N_{FFT} / 2))$$

Lab Measurements: Scopes and receivers

- Purely electrical:
 - Scope + RF Generator: SNR at select sensitivities → FNOB
- Optical and electrical
 - Scope + Receiver dark-noise vs. Sensitivity
 - This returns the setting below which there will be no improvement in SNR
 - Comparison with 50 Ω Termination
 - Scope + Receiver + 1 GHz interference from two RIO ORION lasers
 - SNR: Adjust LO level at fixed “signal” power
 - SNR: Adjust “signal” power at fixed LO power
- By measuring SNR and fraction of scope full-scale, we can calculate the frequency-domain number of bits, “FNOB”



Scope comparison: Electrical (1.05 GHz)

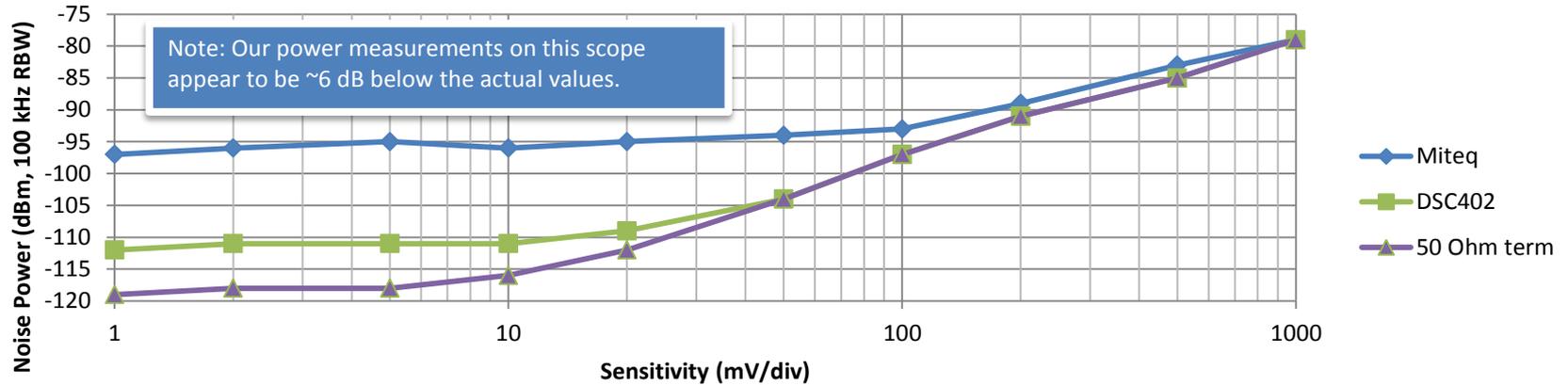


“FNOB” = $(1/6.02) * (SNR_{f,dB} - 1.76 - 20 * \log(2 * RMS / V_{FS}) - 10 * \log(N_{FFT} / 2))$

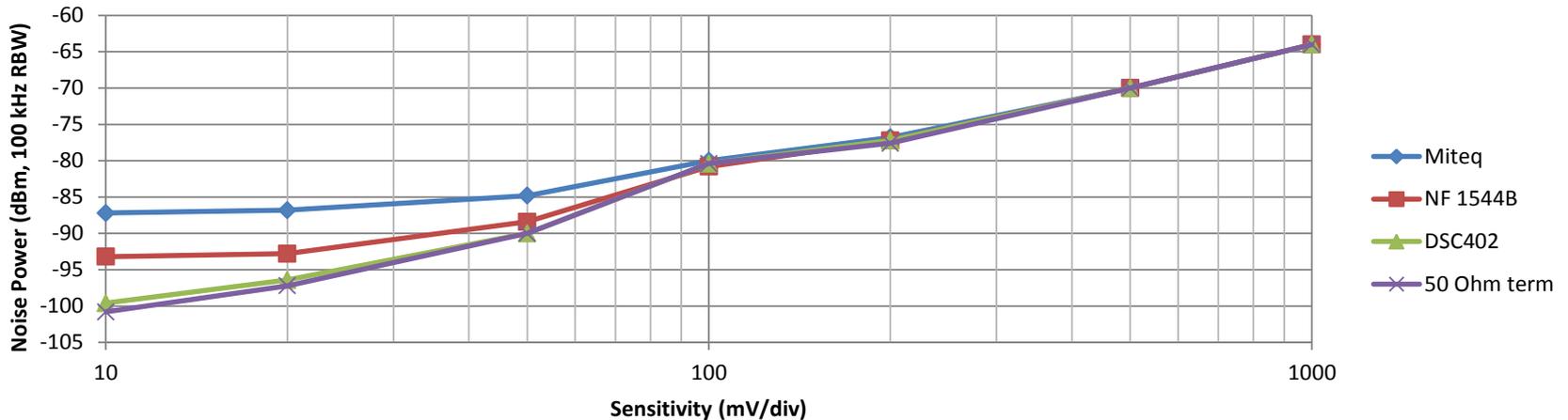


Scope Comparison: Optical Receiver Dark Noise

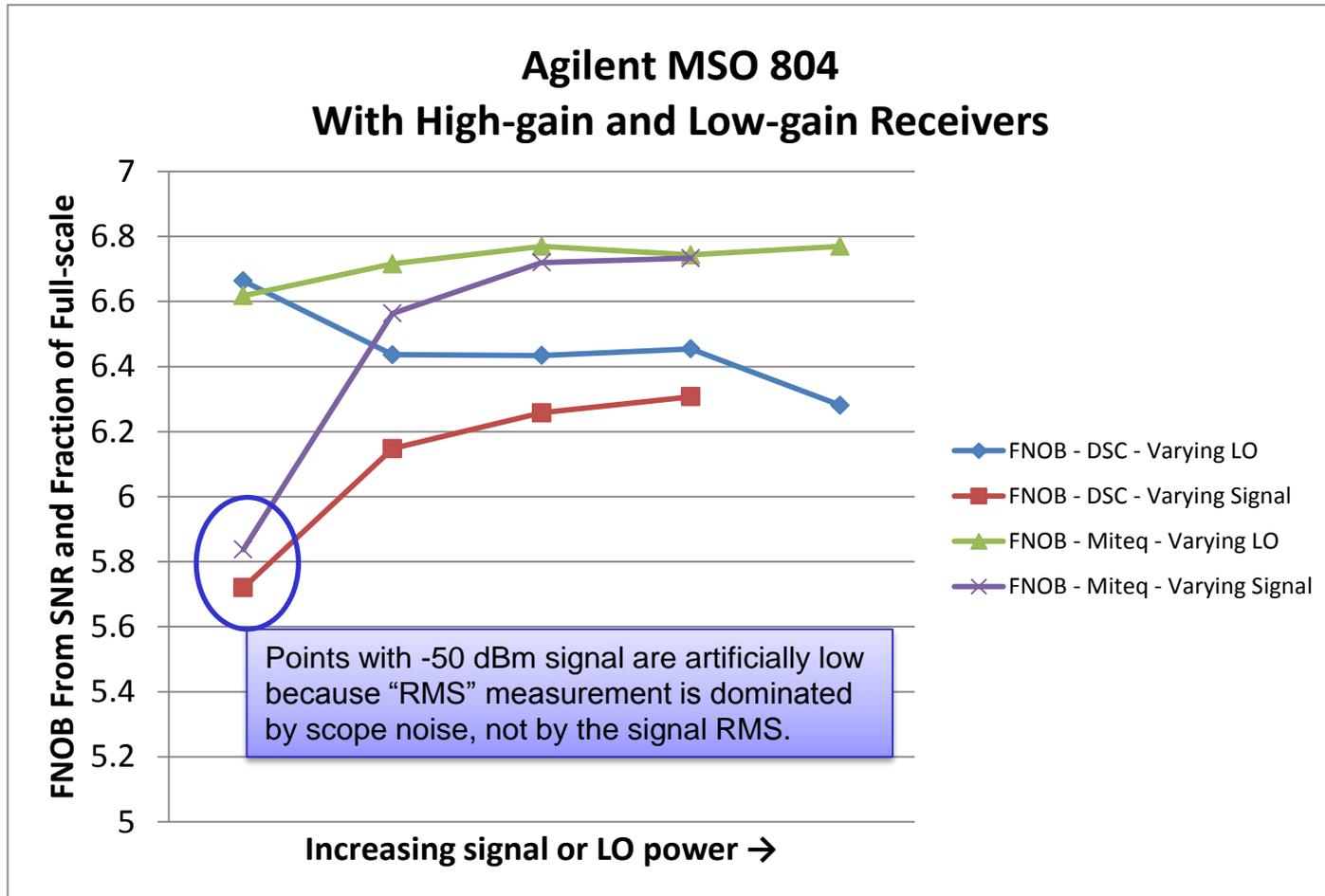
Agilent MSO 804 / Receiver Dark Noise



Tektronix DPO20004 / Receiver Dark Noise



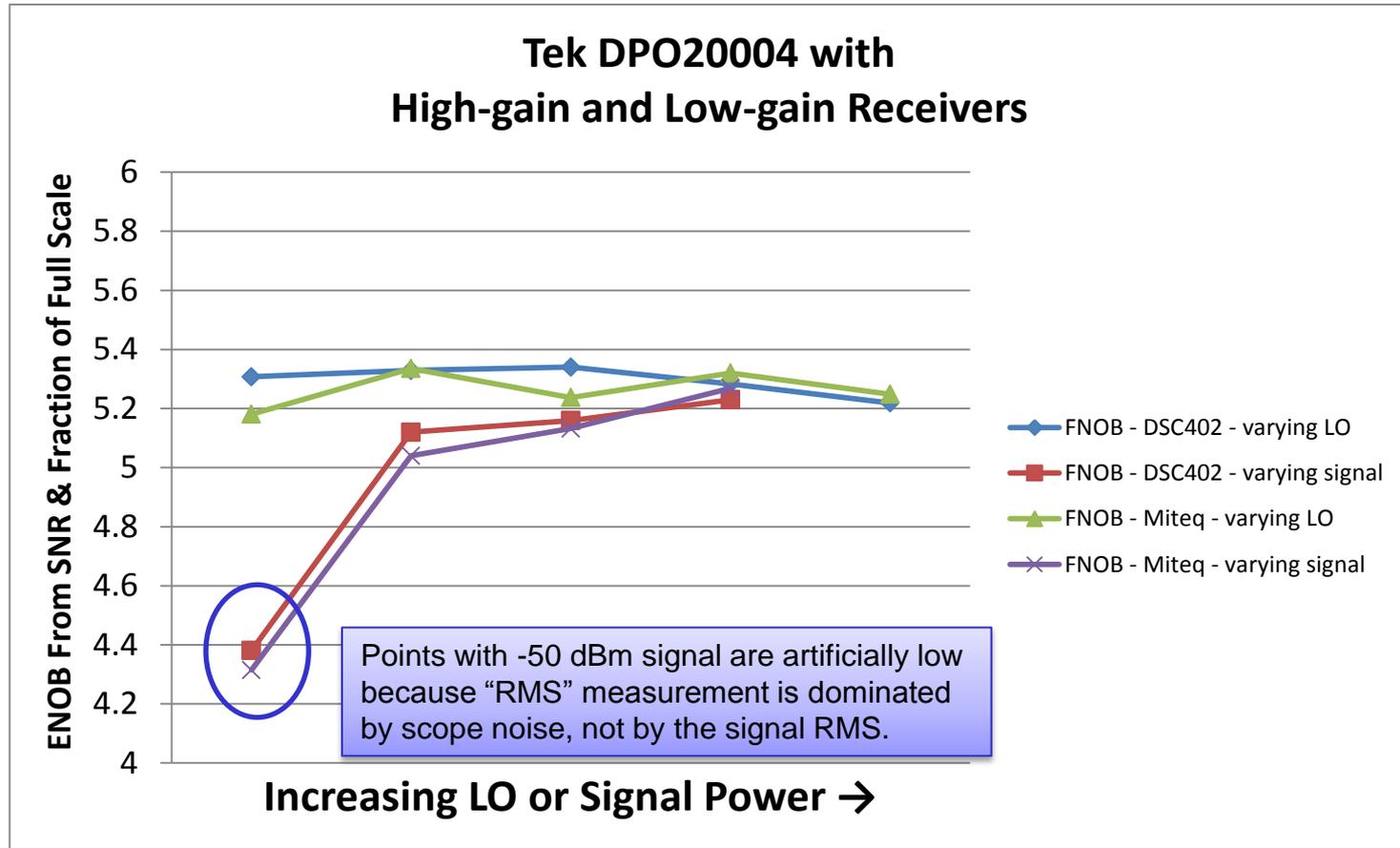
Lab Data: Optical signals on the Agilent MSO 804



Varying LO: -20 dBm to 0 dBm in steps of 5 dBm; Signal = -20 dBm
 Varying Signal: -50 dBm to -20 dBm in steps of 10 dBm; LO = 0 dBm



Lab Data: Optical signals on the Tektronix DPO20004



Varying LO: -20 dBm to 0 dBm in steps of 5 dBm; Signal = -20 dBm
 Varying Signal: -50 dBm to -20 dBm in steps of 10 dBm; LO = 0 dBm



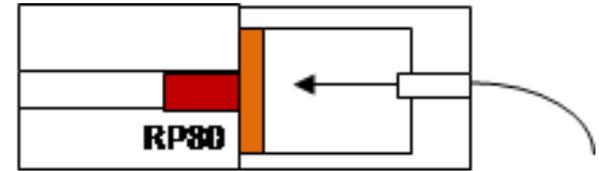
Lab data: Conclusions

LAB DATA	Electrical FNOB	Optical FNOB
Agilent MSO 804	7.4 ± 0.2	6.5 ± 0.3
Tek DPO20004	5.8 ± 0.1	5.25 ± 0.1

- Idealized conditions:
 - No circulators
 - No long transmission fibers
 - No back-reflections
- Agilent MSO 804 seems to give higher dynamic range
 - Can we exploit this on shot data?
- Comparable performance with both receiver configurations
 - Discovery 402 (0.35 V/W) at 10 mV/div
 - Miteq DR125G (2 V/W) at 100 mV/div



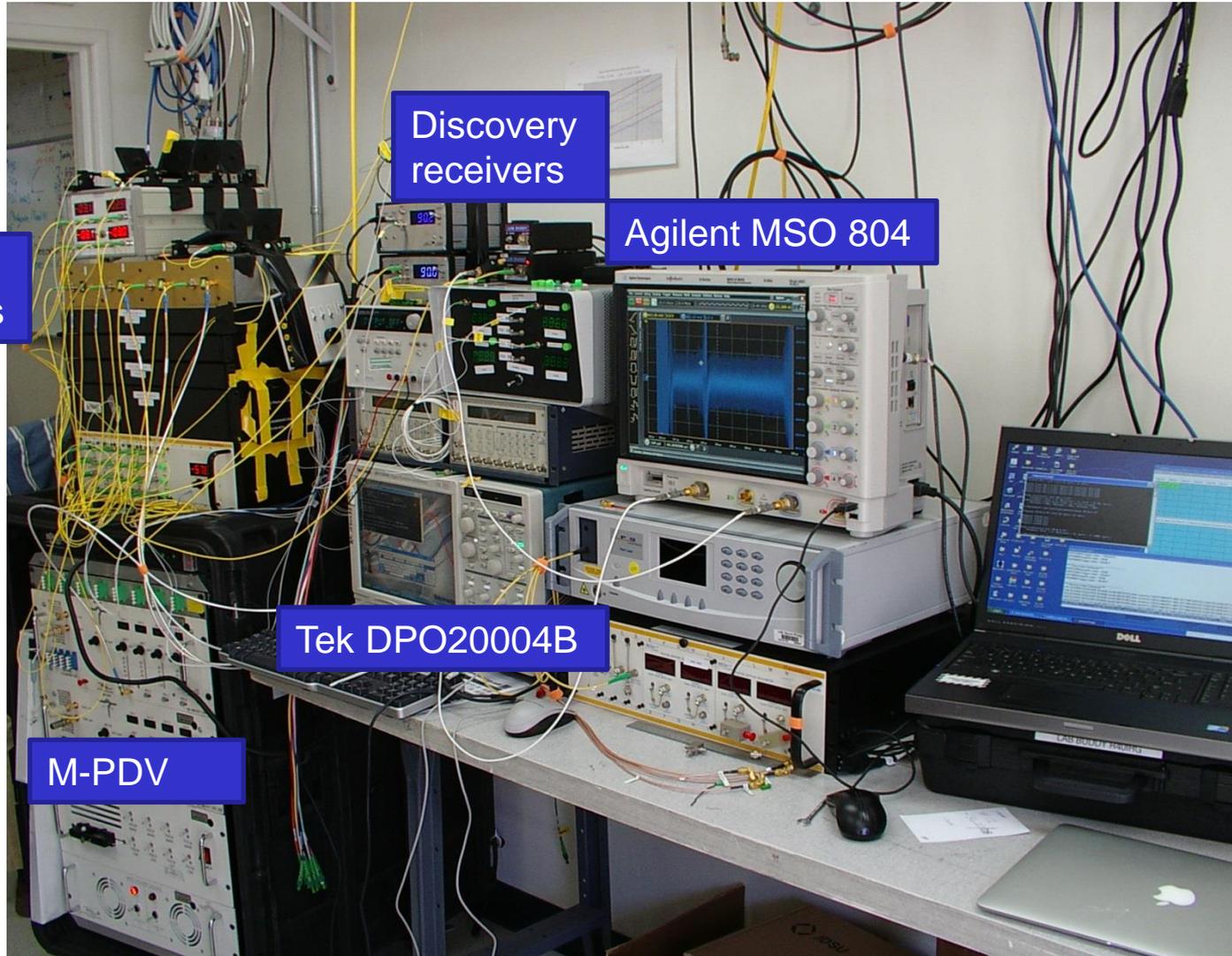
Shot data



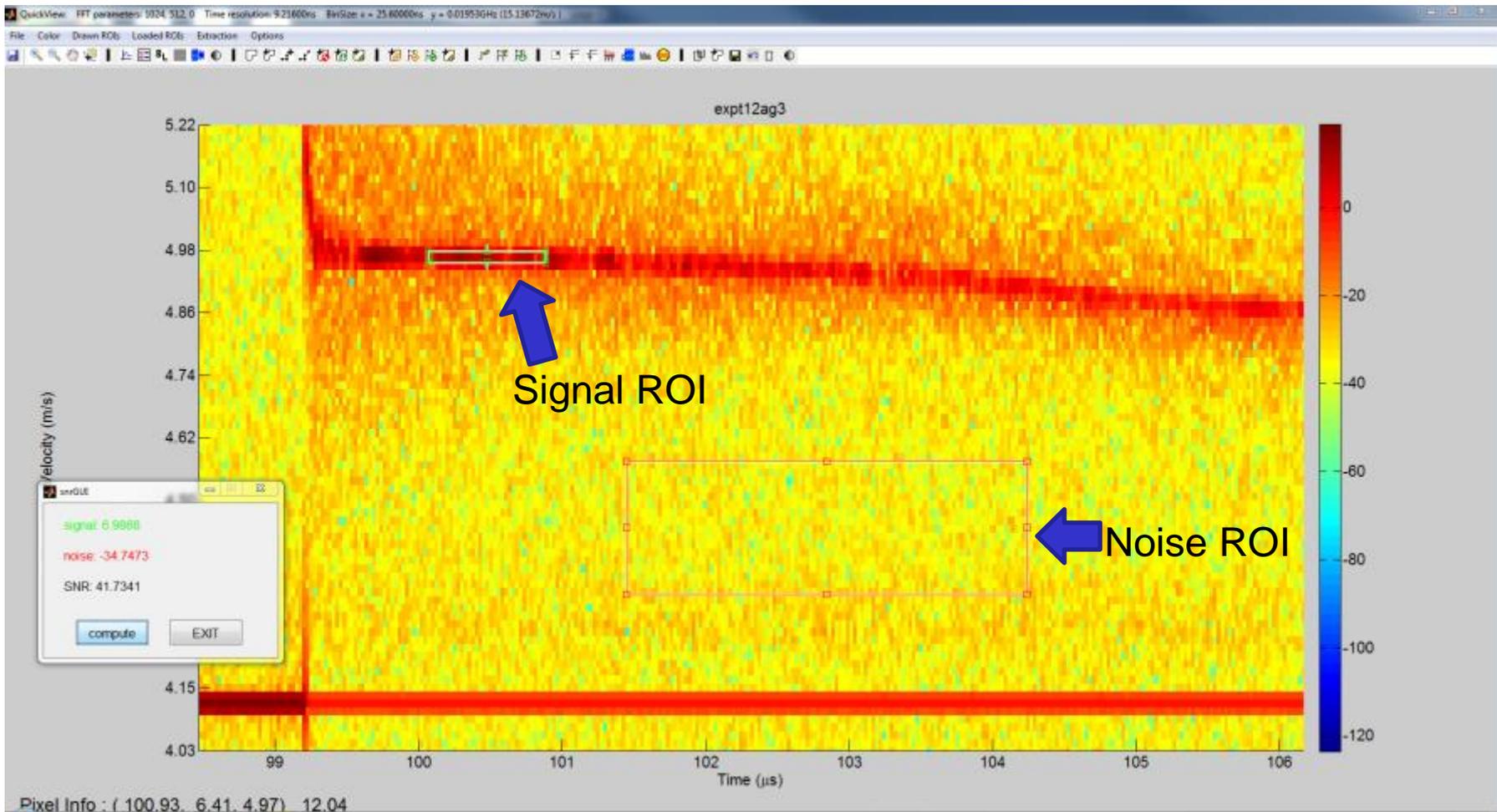
- STL BoomBox
- RP-80 detonators driving 1" diameter copper coupons
- Single PDV probe with multiple wavelengths, demux on back-end
- Velocities ~ 800 m/s
- All systems are heterodyne, zero-velocity frequencies from 1 – 6 GHz
- Variety of recording setups were tested
- Analysis using “Quick-view” with FFT window of 1024 points
- Two scopes used:
 - Tek DPO72004B (run at 25 GS/s, 12.5 GHz)
 - Agilent MSO 804 (run at 20 GS/s, 8.4 GHz)
- Two receivers used:
 - Miteq DR125G (12.5 GHz), ~2 V/mW
 - Discovery DSC402, ~0.35 V/mW



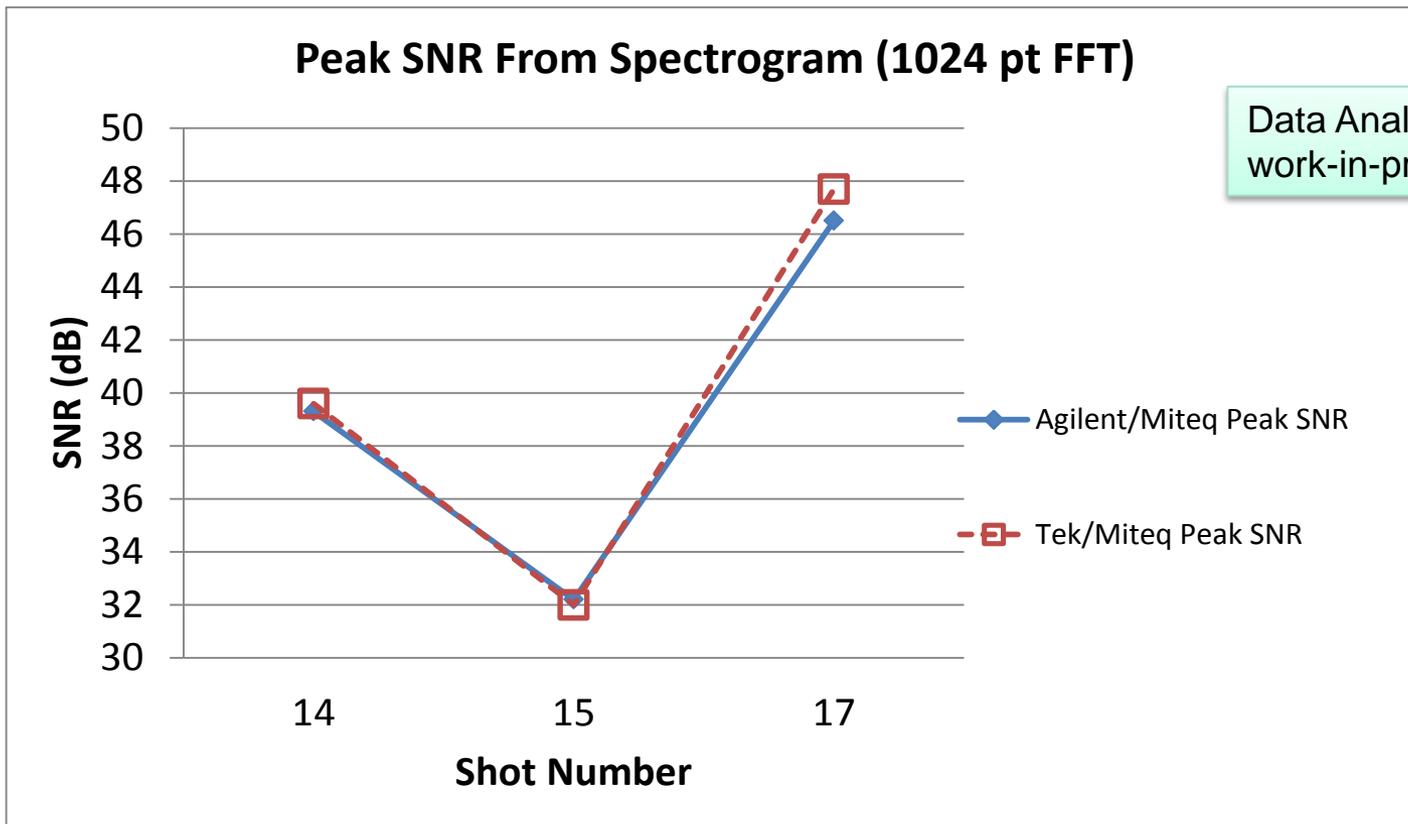
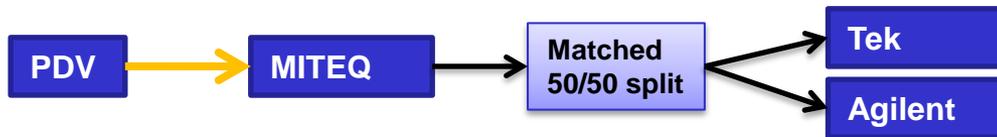
Recording setup



Typical Spectrogram and Analysis



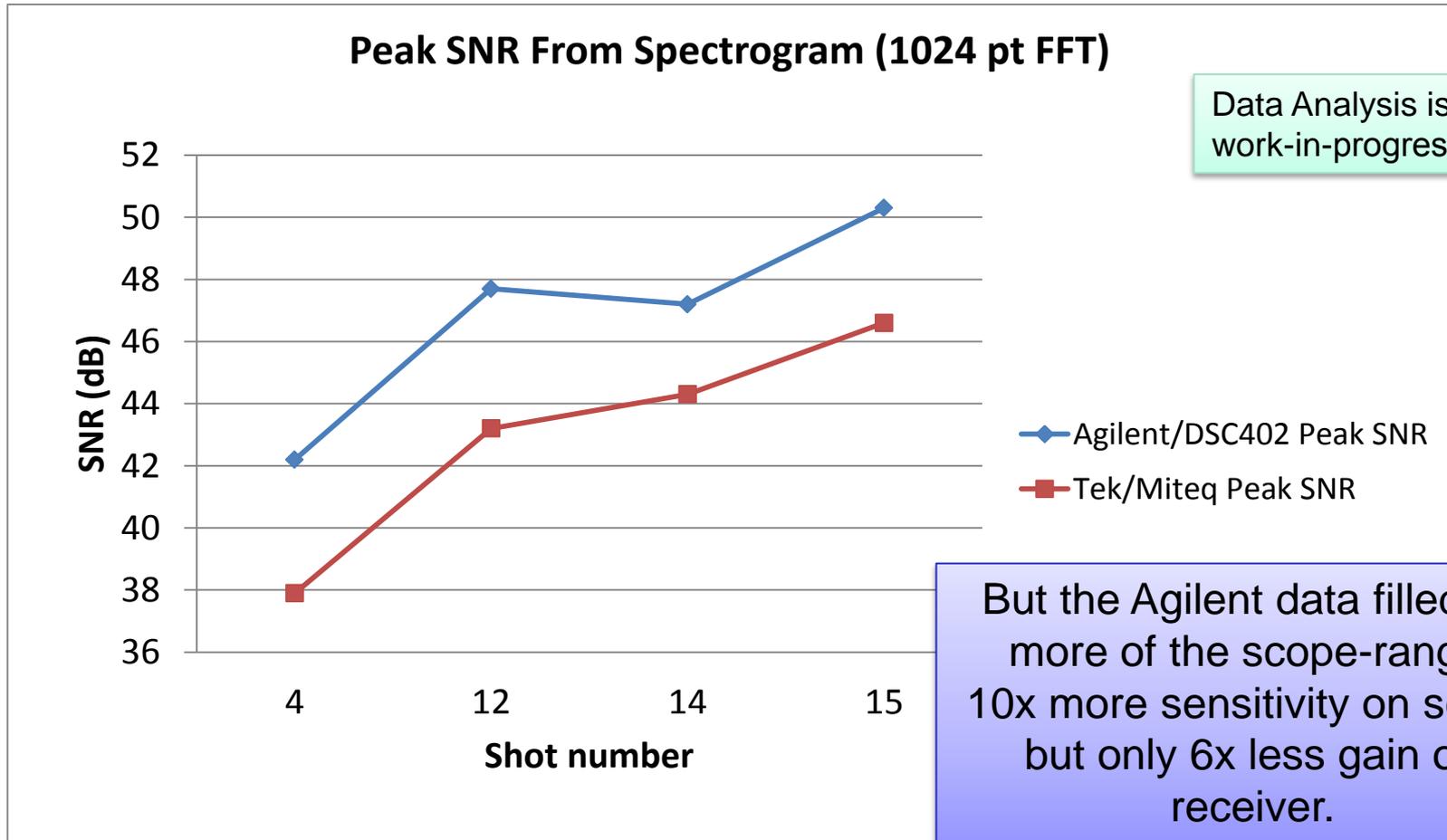
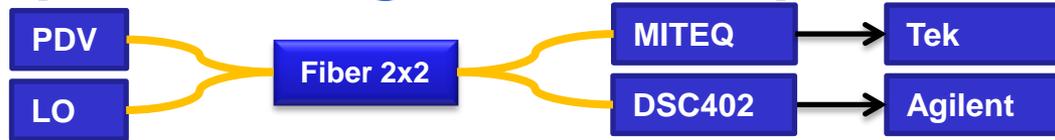
Shot data: Electrical split, Miteq, Agilent and Tek, both on 100 mV/div



Data Analysis is work-in-progress....



Shot data: Optical split, DSC/Agilent & Miteq/Tek



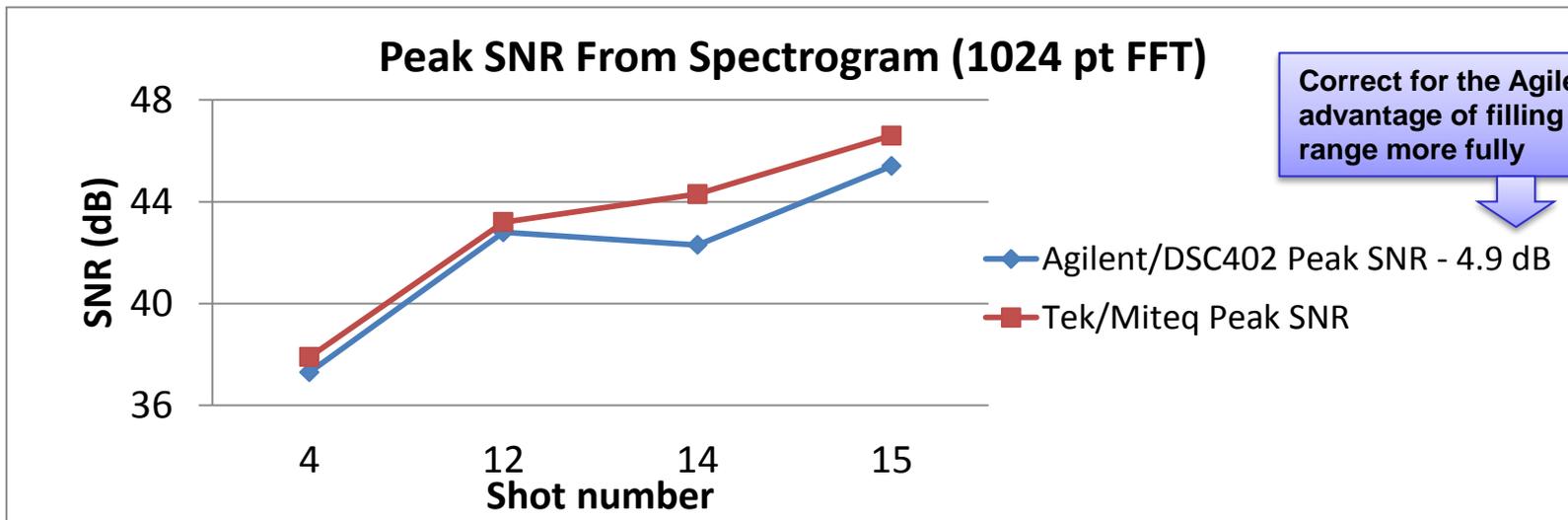
Normalize for scope range

$$SNR_{f,dB} = 6.02 * FNOB + 1.76 + 20 * \log(2 * RMS / V_{FS}) + 10 * \log(N_{FFT} / 2)$$

This will be proportional to
(Receiver Gain) / (Scope Sensitivity)

DSC / Agilent: $20 \log(0.35 \text{ V/mW} / 0.1 \text{ mV}) = 10.9$
 Miteq / Tek: $20 \log(2.00 \text{ V/mW} / 1.0 \text{ V}) = 6.02$

So, just based on fraction of full-scale, the DSC/Agilent setup is getting a bonus 4.9 dB of SNR
 Correct for coverage difference, and the result will be purely the "FNOB" difference



Correct for the Agilent/DSC advantage of filling the Scope range more fully



Conclusions

- Shot data:
 - No clear difference on coverage-normalized shot data
 - With small signals, the DSC receiver will fill up more of the full-scale range at 10 mV/div... potential for better SNR
- Outstanding questions
 - Why does the dynamic range on actual shot data seem to be limited by the PDV hardware?
 - Circulator leakage?
 - Brillouin backscatter?
 - Something else?
 - What about the tin shots, where there was no distinct surface?
 - What conditions (system design, experiment...) would allow us to exploit the extra dynamic range we see in the lab?
- Shot-data experiments I'd like to do in the future:
 - DSC receiver vs Miteq receiver on the same scope
 - Homodyne PDV – DC-block the zero-velocity signal
 - Put circulator near probe to reduce Rayleigh backscatter



Thank you

- Agilent for loaning the MSO 804 for this work
- NSTec Program Support
 - Detectors & Instrumentation Project
 - Shock-wave R&D Project
- LANL team
- NSTec STL team
- NSTec NLV team
- All the vendors who are making enabling components for PDV systems!

- *And of course, to Ed and Larry for making this workshop possible!*

